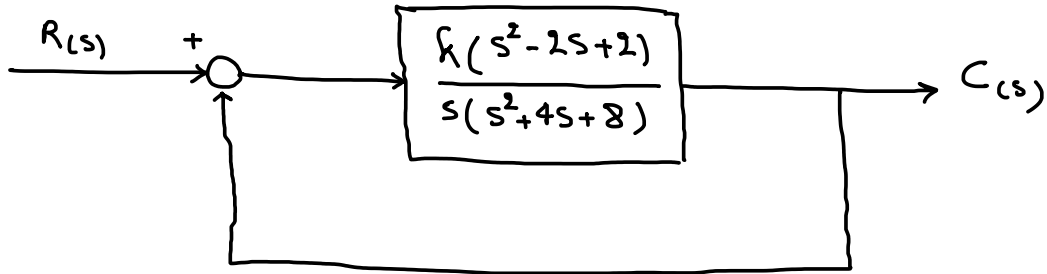


- consider the negative feedback system shown in figure:



Assume that the value of gain  $k$  is nonnegative ( $k \geq 0$ ).

Sketch the root locus plot.

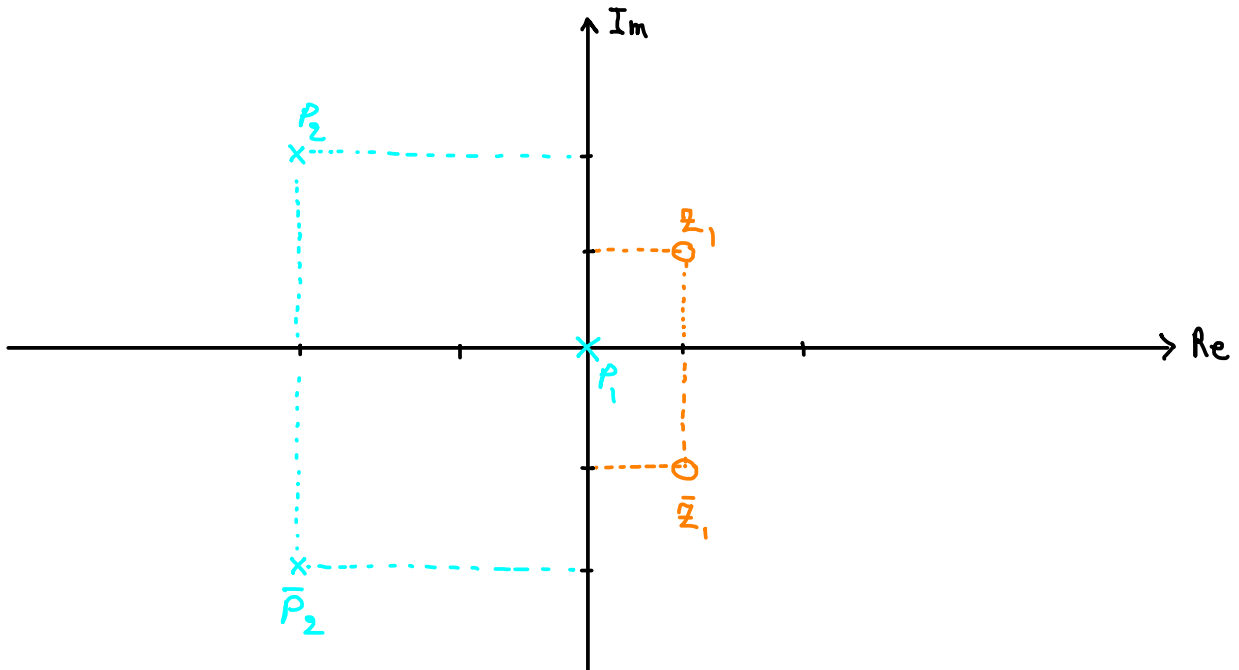
1

1.1 calculate zeros and poles of open-loop transfer function.

Zeros:  $s = \frac{-(-2) \pm \sqrt{(-2)^2 - 4 \times 1 \times 2}}{2 \times 1} = +1 \pm i$  or  $+1 \pm j$       $i = j = \sqrt{-1}$

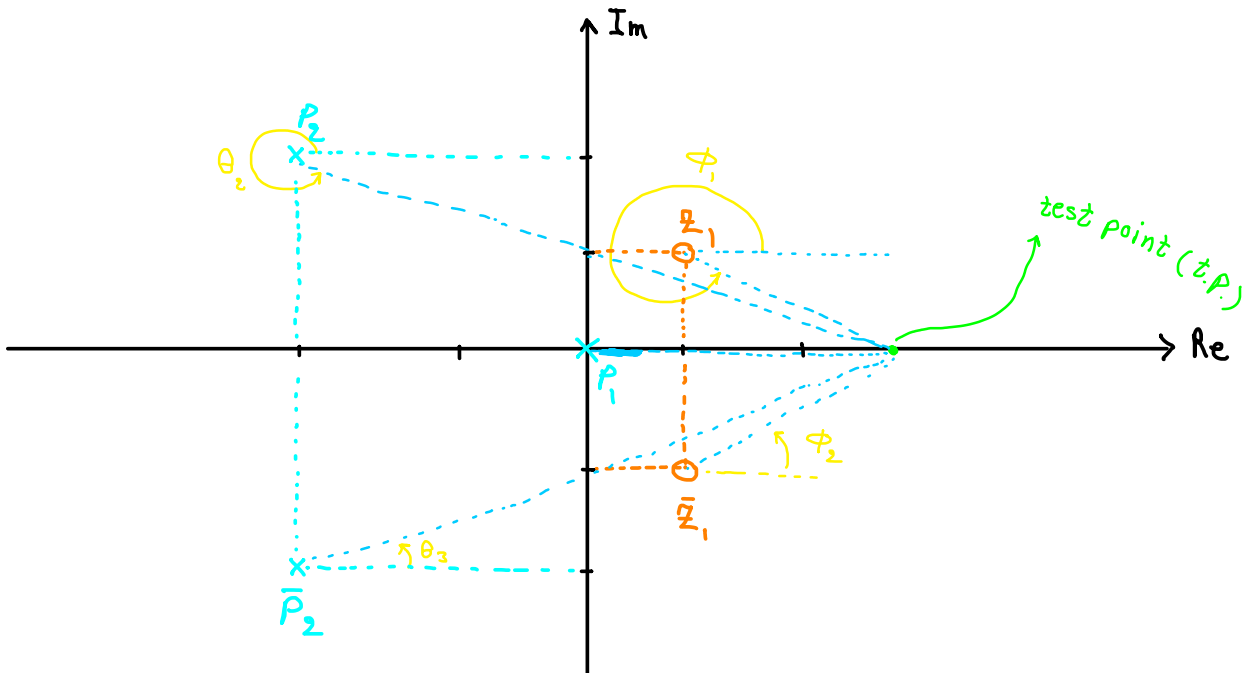
Poles:  $s = 0$  and  $s = \frac{4 \pm \sqrt{4^2 - 4 \times 1 \times 8}}{2 \times 1} = -2 \pm 2i$

1.2 locate the zeros and poles of open-loop T.F. on the  $s$  plane.



9) Plot the root locus for the system with the following transfer function.

2 Determine the root loci on the real axis.

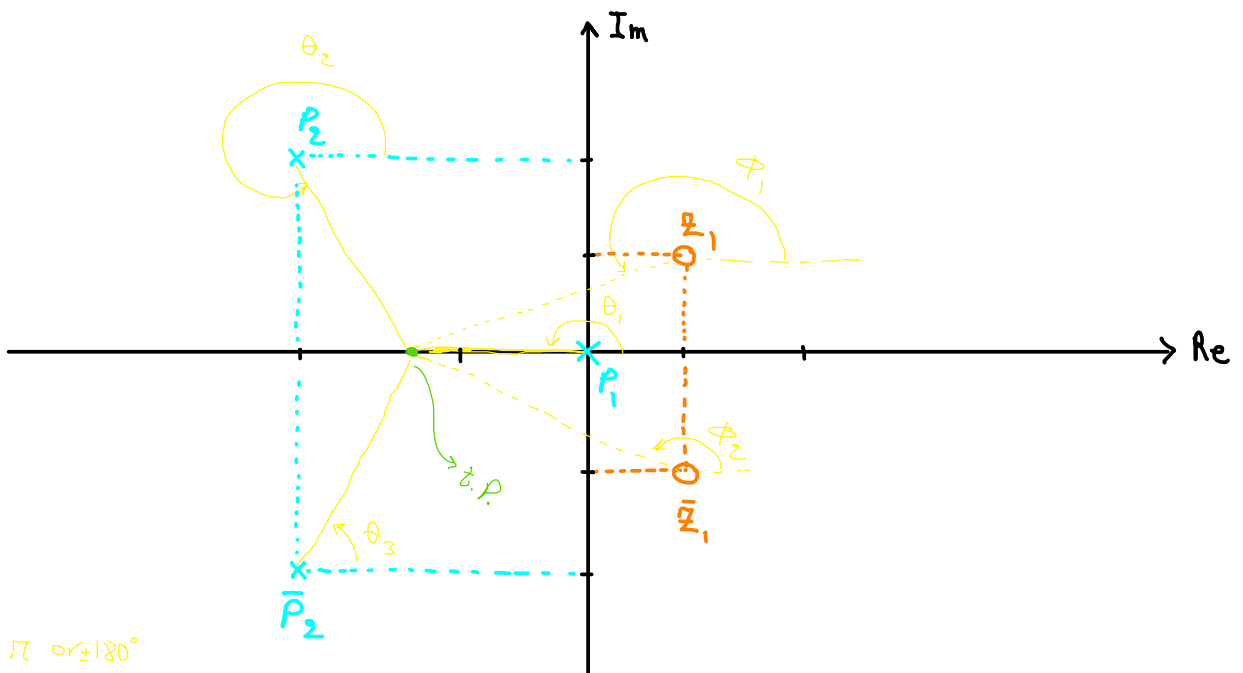


$$\theta_1, \angle P_1 = 0, \quad \phi_1 + \phi_2 = \angle Z_1 + \angle \bar{Z}_1 = 2\pi, \quad \theta_2 + \theta_3 = \angle P_2 + \angle \bar{P}_2 = 2\pi$$

$$\text{Angle condition: } -\theta_1 - (\theta_2 + \theta_3) + (\phi_1 + \phi_2) = -\angle P_1 - (\angle P_2 + \angle \bar{P}_2) + (\angle Z_1 + \angle Z_2) \stackrel{?}{=} \pm(2k-1)\pi$$

$-0 - 2\pi + 2\pi \neq \pm(2k-1)\pi \Rightarrow$  The angle condition is not satisfied.

2.2



$$\theta_1 = \angle P_1 = \pi \text{ or } \pm 180^\circ$$

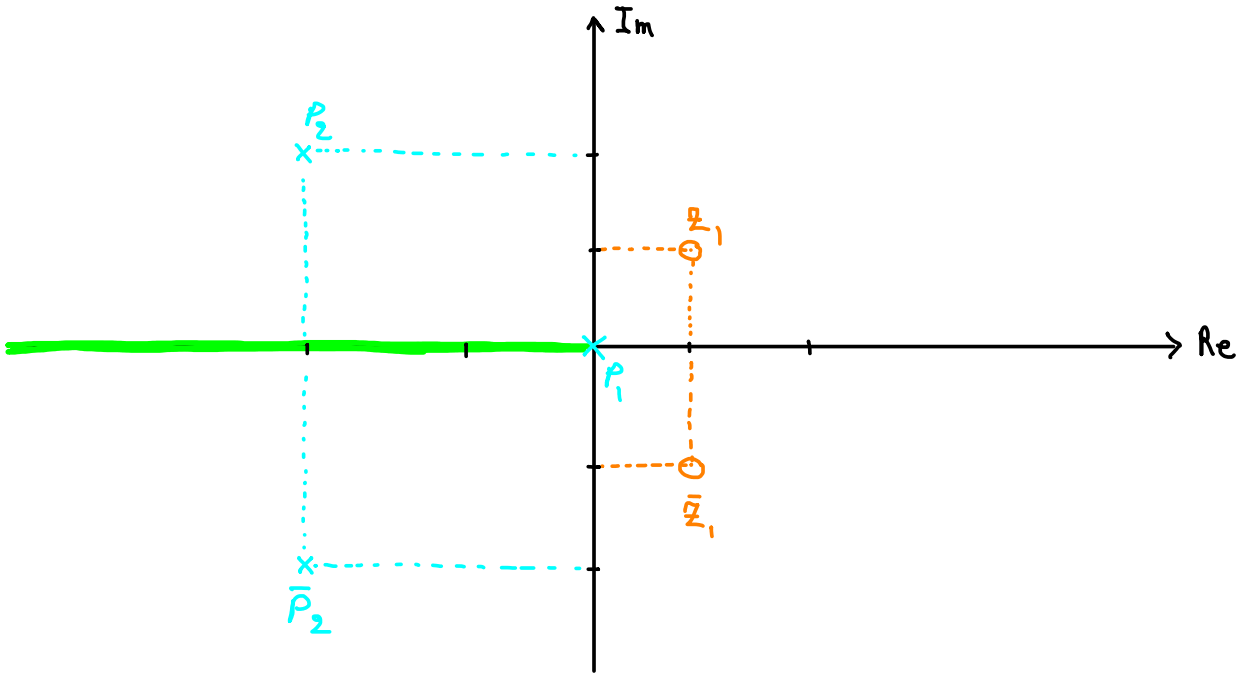
$$\theta_2 + \theta_3 = \angle P_2 + \angle \bar{P}_2 = 2\pi$$

$$\phi_1 + \phi_2 = \angle z_1 + \angle \bar{z}_1 = 2\pi$$

$$\text{Angle condition: } -\theta_1 - (\theta_2 + \theta_3) + (\phi_1 + \phi_2) \stackrel{?}{=} \pm(2k-1)\pi \Rightarrow$$

$$-\pi - (2\pi) + (2\pi) = -\pi = \pm(2k-1)\pi, \quad k=0$$

The angle condition is satisfied.



**3** Determine the asymptots of the root loci

**3.1** Determine number of asymptots.

$$N = n - m = 3 - 2 = 1$$

$$N = 1$$

**3.2** Determine the angle of asymptots.

$$\phi_A = \frac{\pm 180^\circ (2k-1)}{n-m} = \frac{\pm 180^\circ (2k-1)}{3-2} = \pm 180^\circ (2k-1), \quad k=0 \Rightarrow \phi_A = \pm 180^\circ$$

**3.3** The asymptots intersect at a point on the real axis.

$$\sigma_A = \frac{\sum \text{Re}(p_i) - \sum \text{Re}(z_j)}{n-m} = \frac{(0-2-2) - (1+1)}{3-2} = -6$$

4 Find the breakaway and break-in points.

4.1 Write the characteristic equation.

$$1 + \frac{k(s^2 - 2s + 2)}{s(s^2 + 4s + 8)} = 0 \quad (1)$$

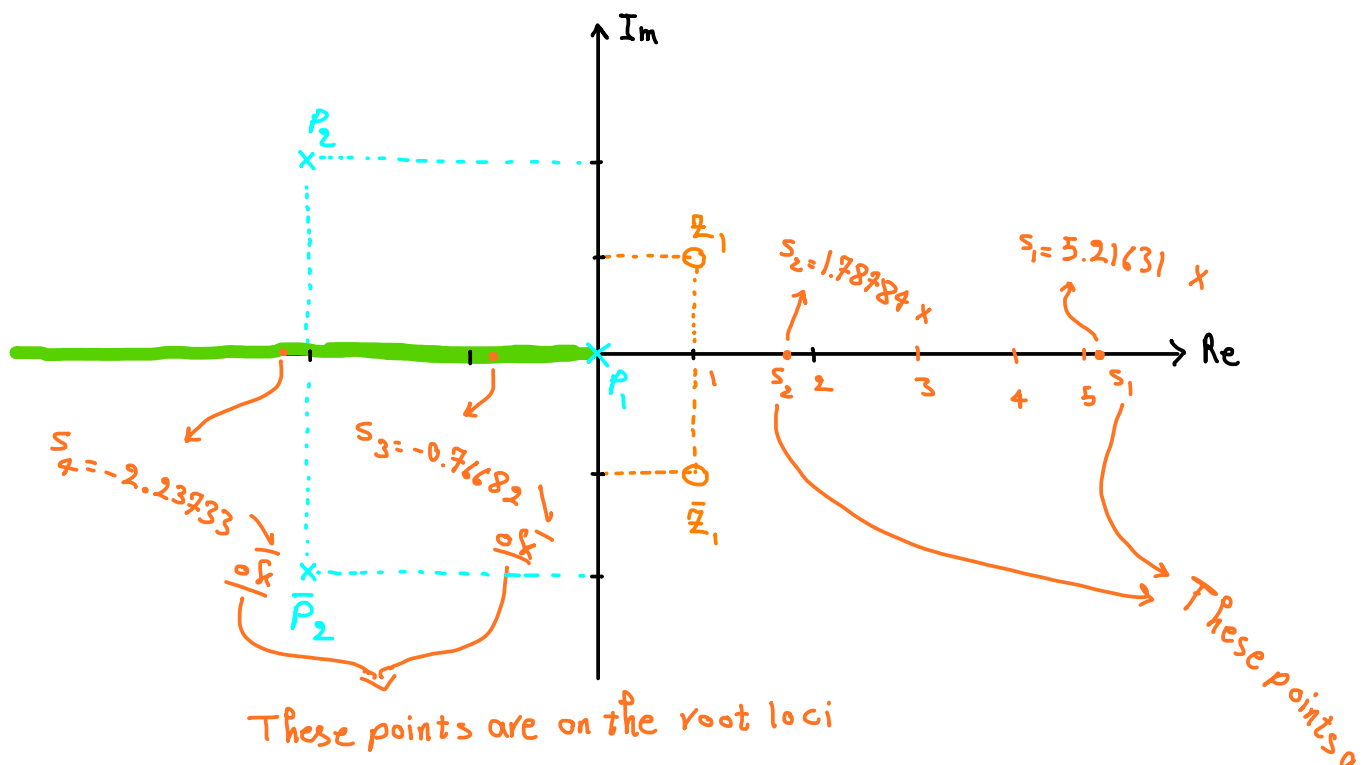
4.2 Obtain  $k$  from the characteristic equation.

$$(1) \Rightarrow \frac{k(s^2 - 2s + 2)}{s(s^2 + 4s + 8)} = -1 \Rightarrow k = \frac{s(s^2 + 4s + 8)}{s^2 - 2s + 2} = \frac{s^3 + 4s^2 + 8s}{s^2 - 2s + 2} \quad (2)$$

4.3 Determine  $s$  from  $\frac{dk}{ds} = 0$

$$(2) \Rightarrow \frac{dk}{ds} = \frac{(3s^2 + 8s + 8)(s^2 - 2s + 2) - (2s - 2)(s^3 + 4s^2 + 8s)}{(s^2 - 2s + 2)^2} = \frac{s^4 - 4s^3 - 10s^2 + 16s + 16}{(s^2 - 2s + 2)^2}$$

$$\frac{dk}{ds} = 0 \Rightarrow s^4 - 4s^3 - 10s^2 + 16s + 16 = 0 \Rightarrow \begin{cases} s_1 = 5.21631 \\ s_2 = 1.78784 \\ s_3 = -0.76682 \\ s_4 = -2.23733 \end{cases}$$





These points are on the root loci

These points are not on the root loci

- Calculate  $k$  for  $s_1, s_2, s_3$  and  $s_4$ :

$$o \quad s_1 = 5.21631 \xrightarrow{(2)} k|_{s=s_1} = \frac{s_1^3 + 4s_1^2 + 8s_1}{s_1^2 - 2s_1 + 2} = -15.5776 < 0 \quad \times$$

$$o \quad s_2 = 1.78784 \xrightarrow{(2)} k|_{s=s_2} = \frac{s_2^3 + 4s_2^2 + 8s_2}{s_2^2 - 2s_2 + 2} = -20.2400 < 0 \quad \times$$

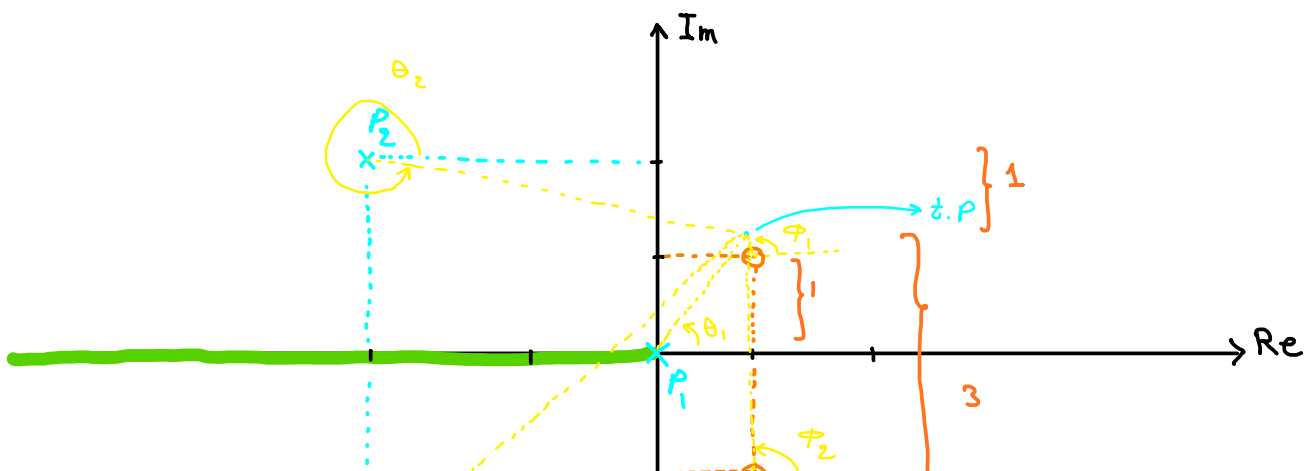
$$o \quad s_3 = -0.76682 \xrightarrow{(2)} k|_{s=s_3} = \frac{s_3^3 + 4s_3^2 + 8s_3}{s_3^2 - 2s_3 + 2} = 1.0271 \geq 0 \quad \checkmark \text{ ok}$$

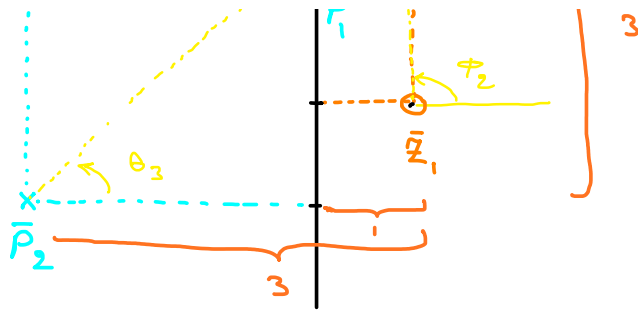
$$o \quad s_4 = -2.23733 \xrightarrow{(2)} k|_{s=s_4} = \frac{s_4^3 + 4s_4^2 + 8s_4}{s_4^2 - 2s_4 + 2} = 0.7905 \geq 0 \quad \checkmark \text{ ok}$$

5 Determine the angle of departure (angle of arrival) of the root locus

from a complex poles (at a complex zeros).

5.1





$$\phi_1 = ? \quad \phi_2 \approx 90^\circ \quad \theta_3 \approx \tan^{-1}\left(\frac{3}{3}\right) = 45^\circ, \quad \theta_2 \approx 360^\circ - \tan^{-1}\left(\frac{1}{3}\right) = 360^\circ - 18.435^\circ = 341.565^\circ$$

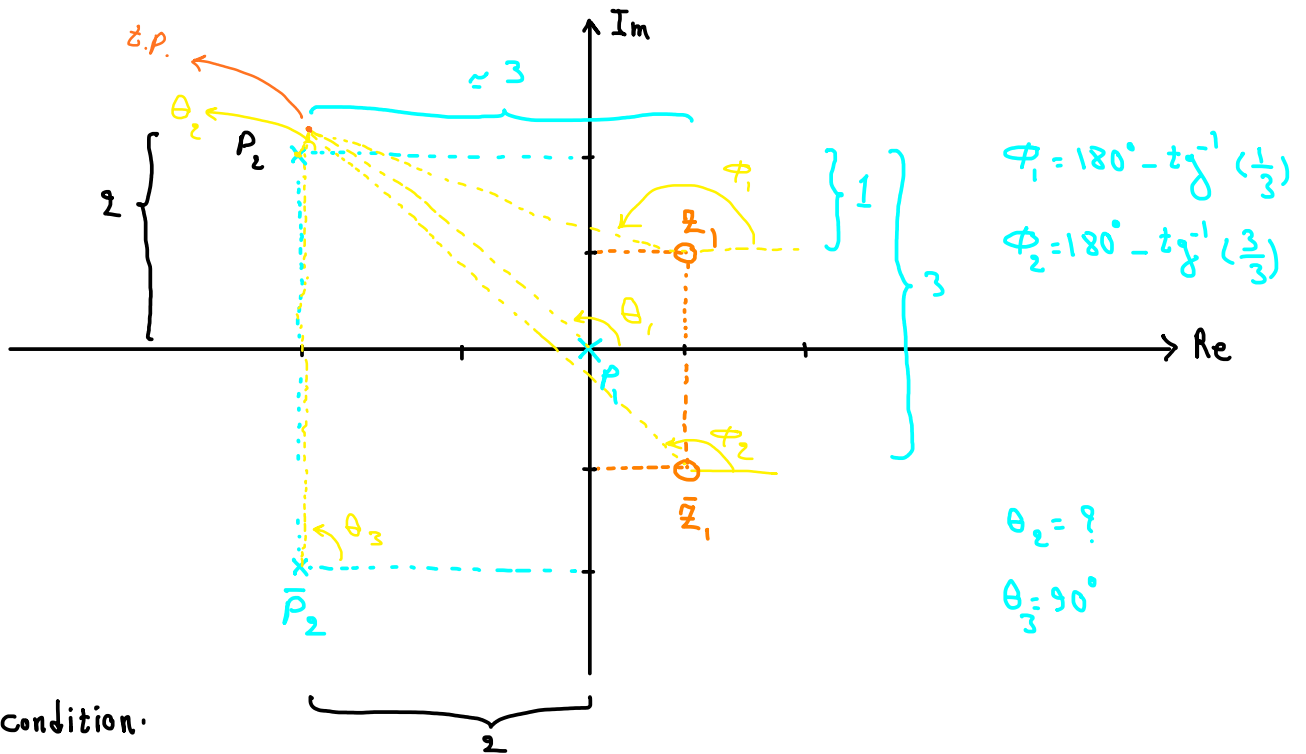
$$\theta_1 = \tan^{-1}\left(\frac{1}{1}\right) = 45^\circ$$

$$\text{Angle condition: } -\theta_1 - \theta_2 - \theta_3 + \phi_1 + \phi_2 = \pm(2k-1)180^\circ \Rightarrow$$

$$-45^\circ - 341.565^\circ - 45^\circ + \phi_1 + 90^\circ = \pm(2k-1)180^\circ \Rightarrow \phi_1 = \pm(2k-1)180^\circ + 341.565^\circ \Rightarrow$$

$$k=0 \Rightarrow \mp 180^\circ + 341.565^\circ = \phi_1 \Rightarrow \boxed{\phi_1 = 161.565^\circ}$$

5.2



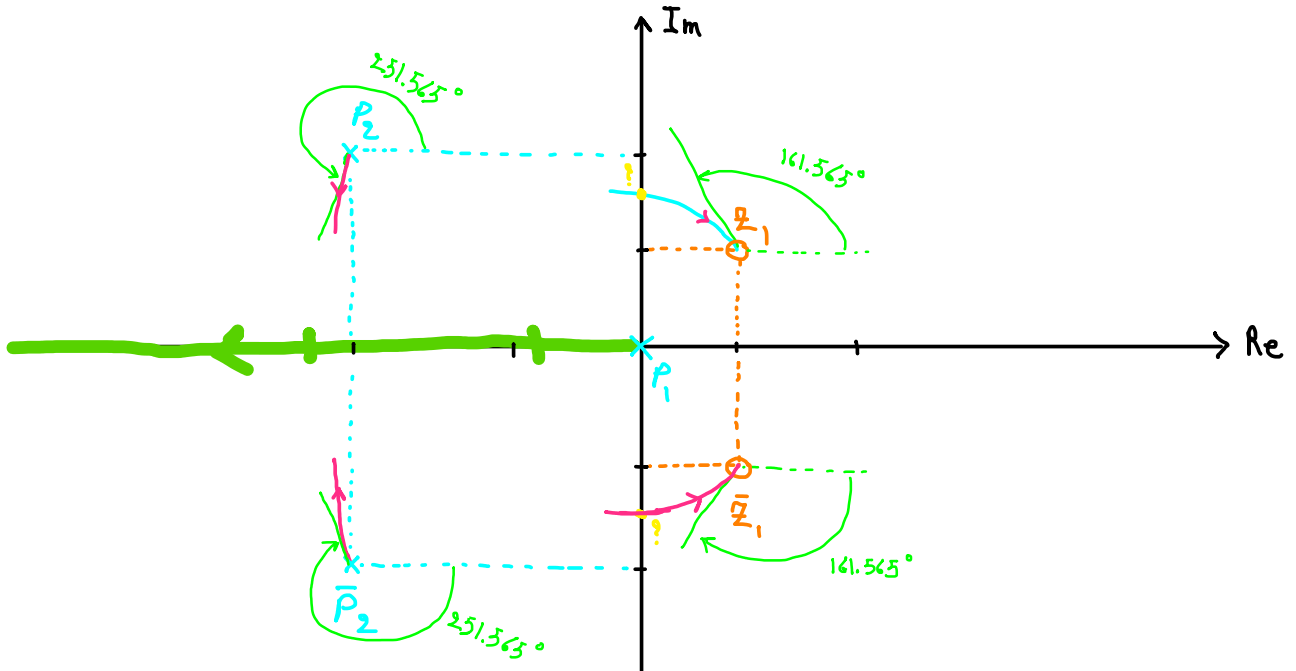
Angle condition.

$$\theta_1 = 180^\circ - \tan^{-1}\left(\frac{2}{2}\right) = 135^\circ, \quad \theta_2 = 90^\circ, \quad \theta_3 = ? \quad \phi_1 = 180^\circ - \tan^{-1}\left(\frac{1}{3}\right) = 161.565^\circ$$

$$\neq 180^\circ - \tan^{-1}\left(\frac{3}{3}\right) = 135^\circ$$

$$-\theta_1 - \theta_2 - \theta_3 + \phi_1 + \phi_2 = \pm (2k-1)180^\circ \Rightarrow -135^\circ - \theta_2 - 90^\circ + 161.565^\circ + 135^\circ = \pm 180^\circ (2k-1)$$

$$\Rightarrow \theta_2 = 71.565^\circ \mp (2k-1)180^\circ \Rightarrow \boxed{\theta_2 = 251.565^\circ}$$



6 Find the points where the root loci cross the imaginary axis.

6.1 Routh's stability criterion

$$(1) \Rightarrow 1 + \frac{k(s^2 - 2s + 2)}{s(s^2 + 4s + 8)} = 0 \Rightarrow \frac{s^3 + 4s^2 + 8s + k(s^2 - 2s + 2)}{s(s^2 + 4s + 8)} = 0 \Rightarrow$$

$$s^3 + (4+k)s^2 + (8-2k)s + 2k = 0$$

$s^3$	1	$(8-2k)$
$s^2$	$(4+k)$	$2k$
$s^1$	$\frac{(4+k)(8-2k) - 1 \times 2k}{(4+k)}$	0
$s^0$	$2k$	

$s^0$  $2k$ 

$$32 - 8k + 8k - 2k^2 - 2k = 0 \Rightarrow k^2 + k - 16 = 0 \Rightarrow$$

$$k = \frac{-1 \pm \sqrt{1^2 - 4 \times 1 \times (-16)}}{2 \times 1} = \begin{cases} k = 3.5311 > 0 \checkmark \\ k = -4.5311 < 0 \times \end{cases}$$

Auxiliary equation:  $(4+k)s^2 + 2k = 0$ ,  $k = 3.5311 \Rightarrow$

$$7.5311s^2 + 7.0622 = 0 \Rightarrow s = \pm \sqrt{\frac{-7.0622}{7.5311}} = \pm 0.968i$$

$$s = \pm 0.968i$$

[6.2] Letting  $s = i\omega = j\omega$  ( $i = j = \sqrt{-1}$ ) in the characteristic equation.

$$s^3 + (4+k)s^2 + (8-2k)s + 2k = 0 \xrightarrow{s=i\omega}$$

$$(i\omega)^3 + (4+k)(i\omega)^2 + (8-2k)(i\omega) + 2k = 0 \Rightarrow$$

$$\underbrace{-i\omega^3}_{\text{Im}} - \underbrace{(4+k)\omega^2}_{\text{Re}} + \underbrace{(8-2k)(i\omega)}_{\text{Im}} + \underbrace{2k}_{\text{Re}} = 0 \Rightarrow$$

$$2k - (4+k)\omega^2 + i[-\omega^3 + (8-2k)\omega] = 0 \Rightarrow \begin{cases} 2k - 4\omega^2 - k\omega^2 = 0 & (*) \\ 8\omega - 2k\omega - \omega^3 = 0 \text{ or } 8 - 2k - \omega^2 = 0 & (**) \end{cases}$$

$$(**) \Rightarrow \omega^2 = 8 - 2k \quad (***)$$

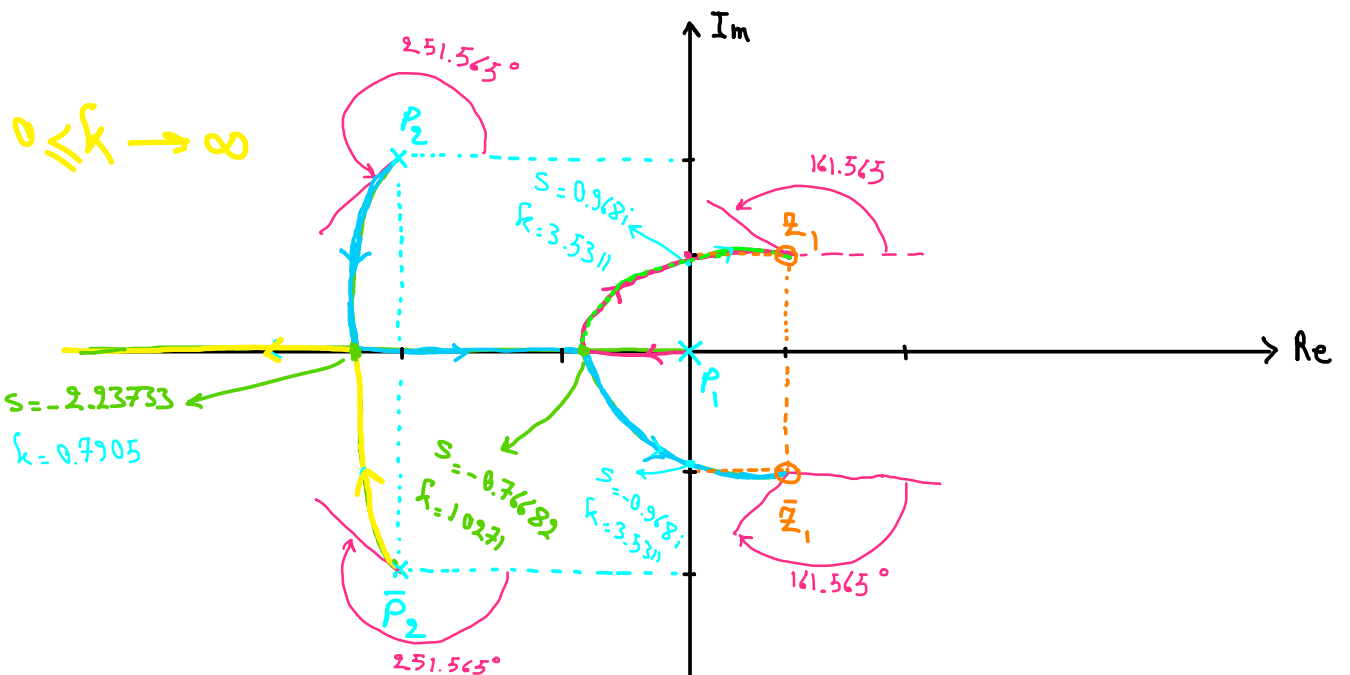
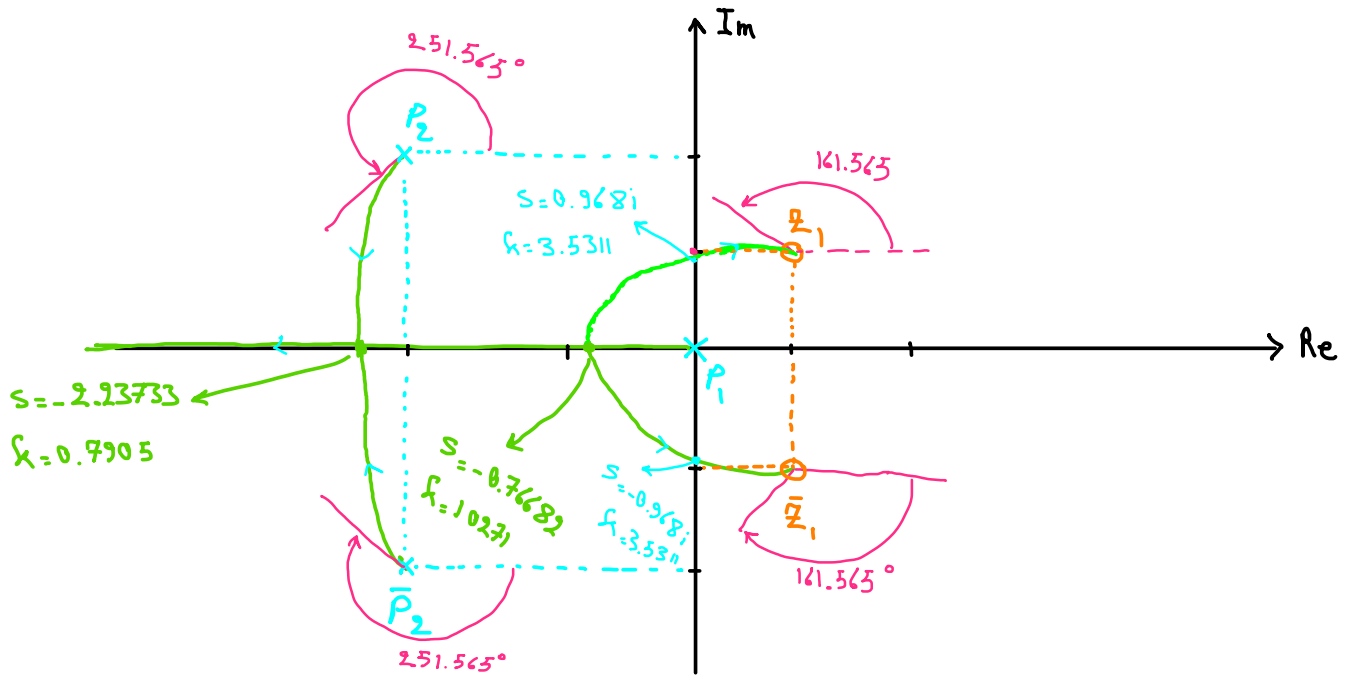
$$(***) \text{ into } (*) \Rightarrow 2k - (4+k)(8-2k) = 0 \Rightarrow 2k^2 + 2k - 32 = 0 \text{ or } \underline{k^2 + k - 16 = 0}$$

$$k = \frac{-1 \pm \sqrt{1^2 - 4 \times 1 \times (-16)}}{2 \times 1} \Rightarrow \begin{cases} k = 3.5311 > 0 \checkmark \\ k = -4.5311 < 0 \times \end{cases}$$

$$(***) \Rightarrow \omega = \pm \sqrt{8 - 2k} \Rightarrow \omega = \pm \sqrt{8 - 2 \times 3.5311} = \pm 0.968 \Rightarrow s = \pm 0.968i$$



$$(***) \Rightarrow \omega = \pm \sqrt{8 - 2k} \Rightarrow \omega = \pm \sqrt{8 - 2 \times 3.5311} = \pm 0.968 \Rightarrow s = \pm 0.968i$$



Example: Sketch bode diagram of  $G_{r(w)} = \frac{k(s^2 + 2s + 2)}{s}$ ,  $k=1$

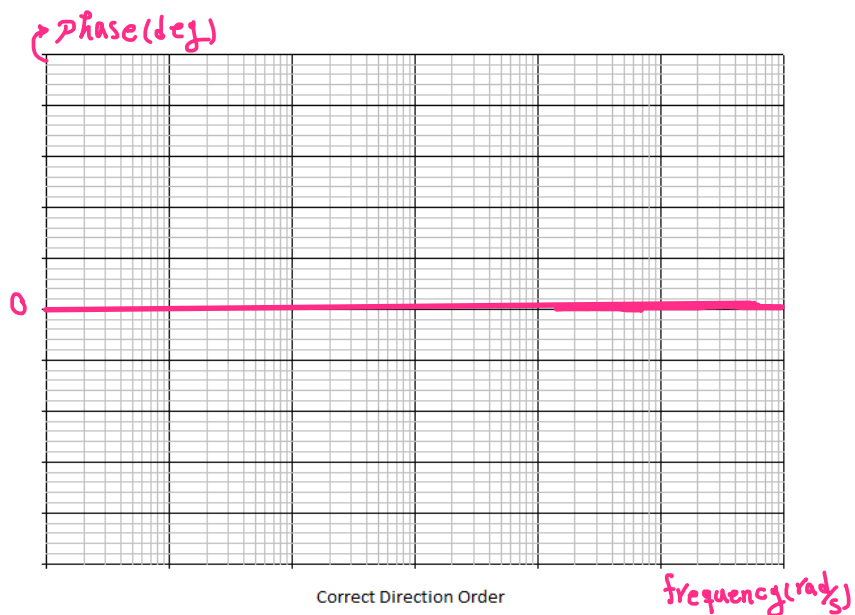
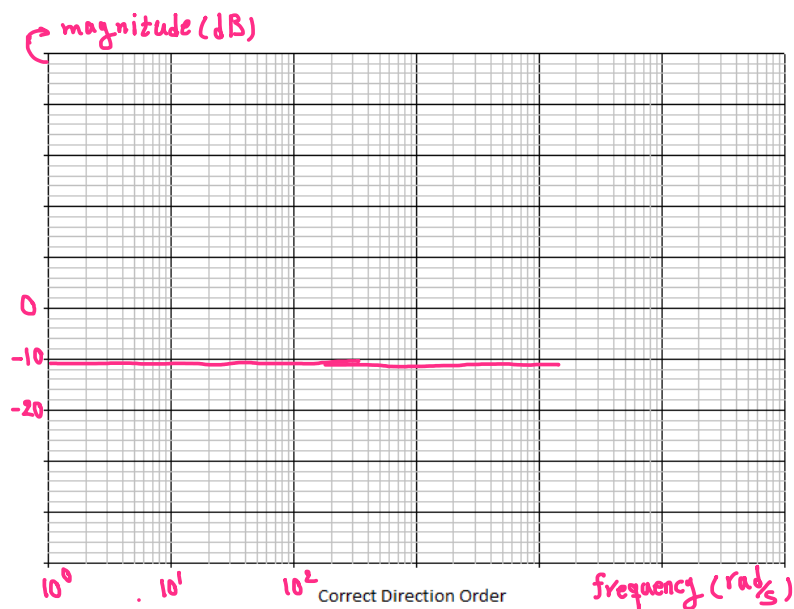
Example: Sketch bode diagram of

$$G_{(j\omega)} = \frac{k(s^2 + 2s + 2)}{s(s^2 + 4s + 8)}, \quad k=1$$

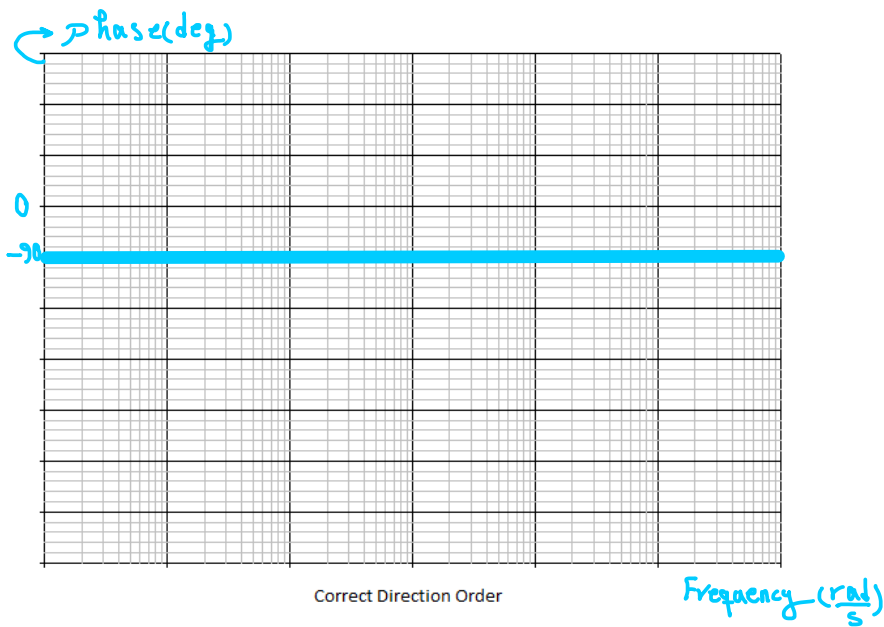
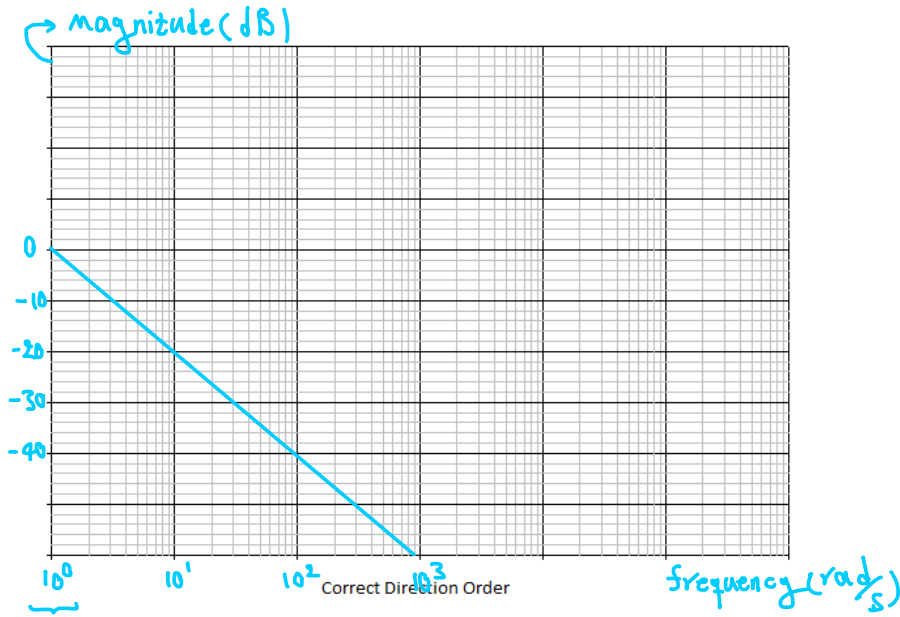
Write the fraction as standard

$$G_{(j\omega)} = \frac{2\left(\frac{s^2}{2} + s + 1\right)}{8\left(\frac{s^2}{8} + \frac{s}{2} + 1\right)} = \frac{2}{8} \frac{\left(1 + s + \frac{s^2}{2}\right)}{\left(1 + \frac{s}{2} + \frac{s^2}{8}\right)}$$

1. The first component =  $\frac{2}{8}$  or  $\frac{1}{4} \Rightarrow 20 \log\left(\frac{1}{4}\right) = -12.0412$

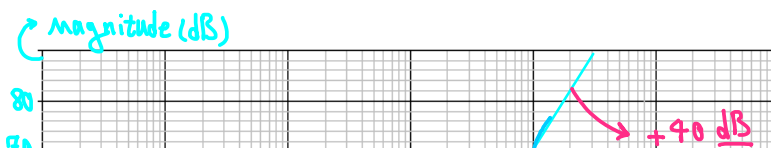


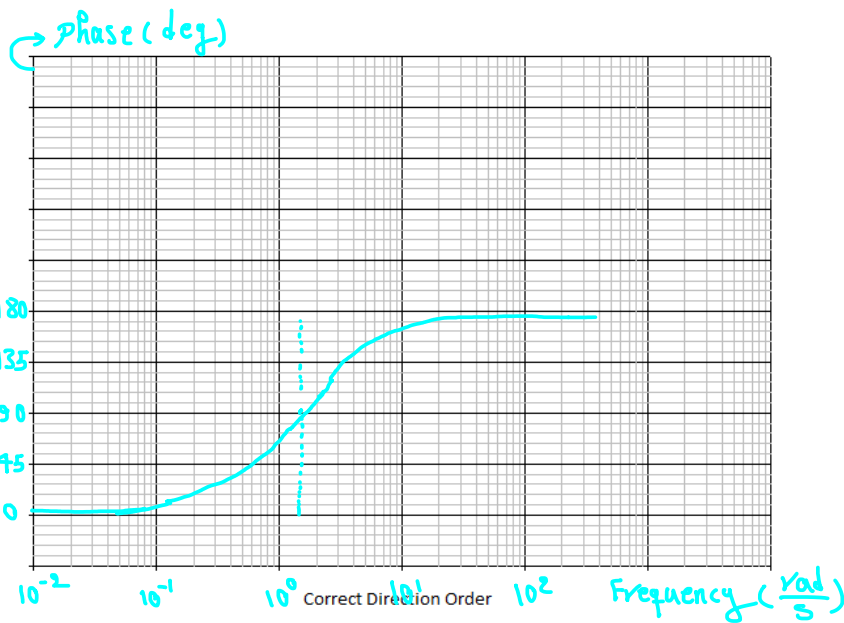
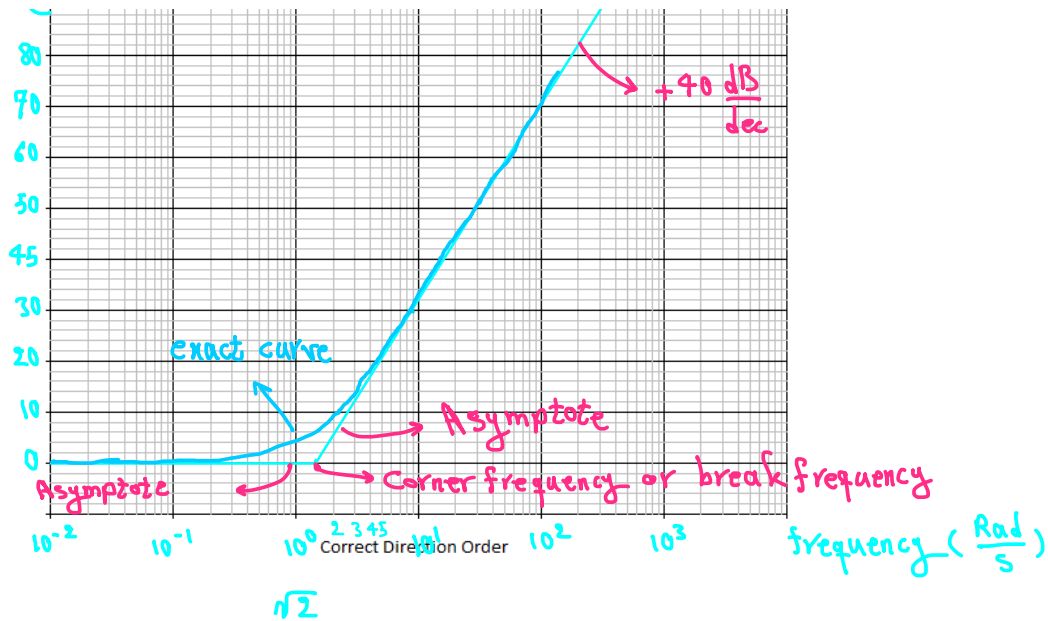
2: The second component:  $\frac{1}{s}$



3) The third component:  $1 + s + \frac{s^2}{2}$  :

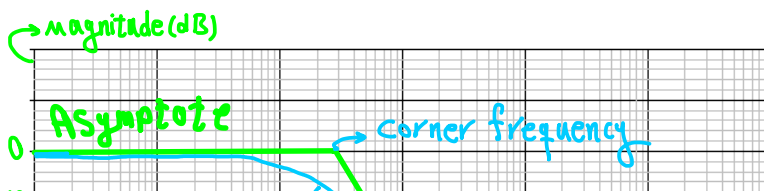
corner frequency or break frequency =  $\sqrt{2}$



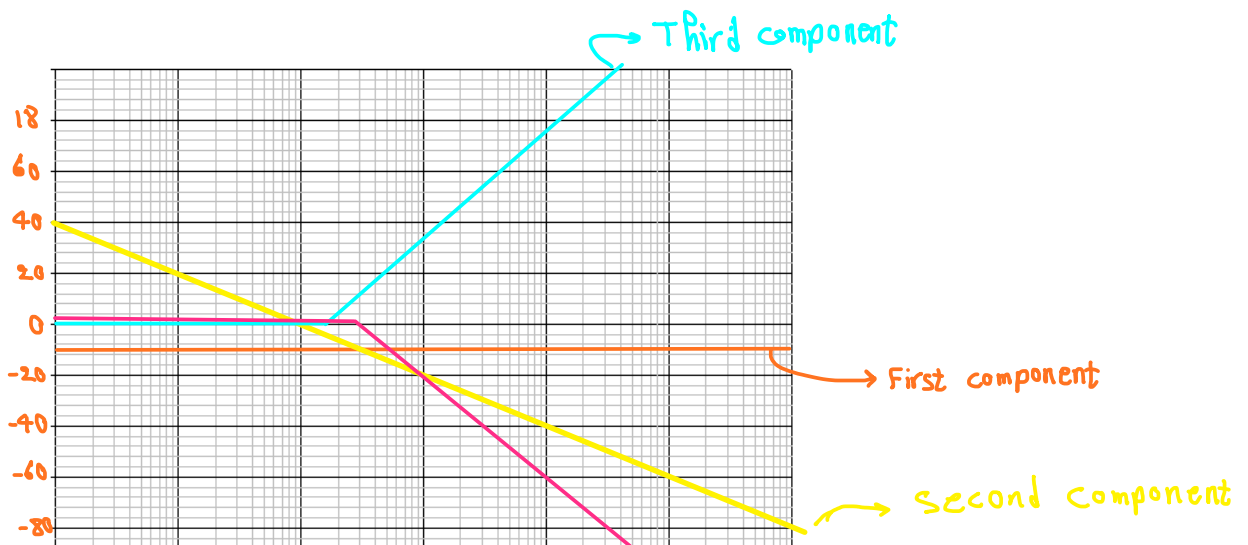
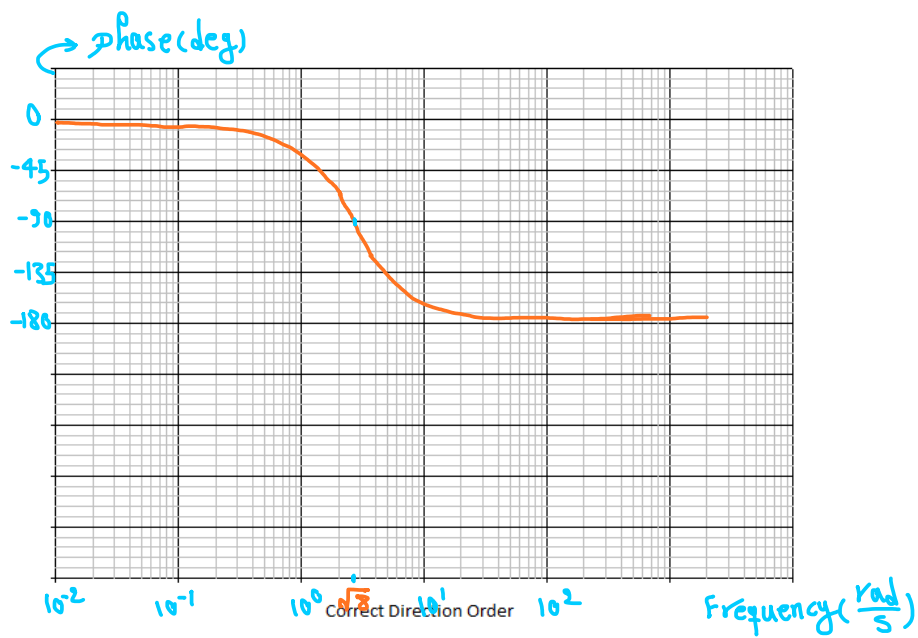
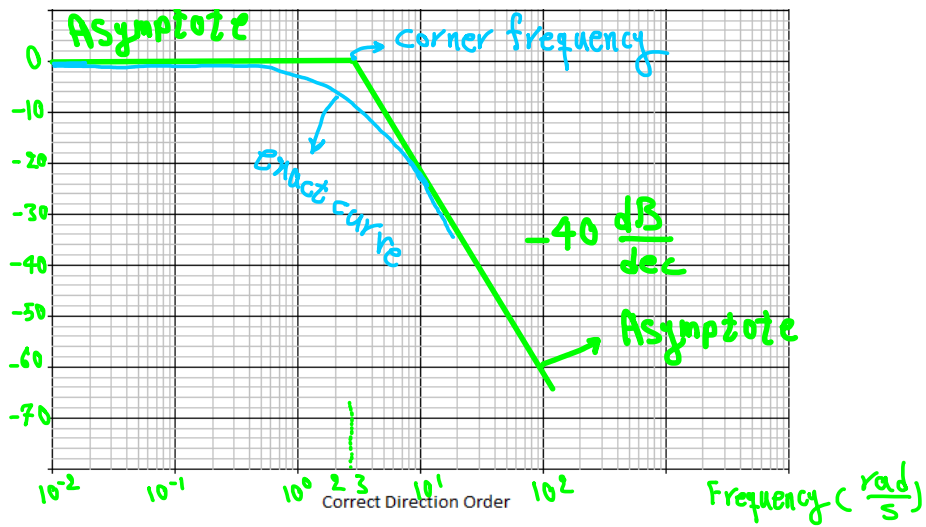


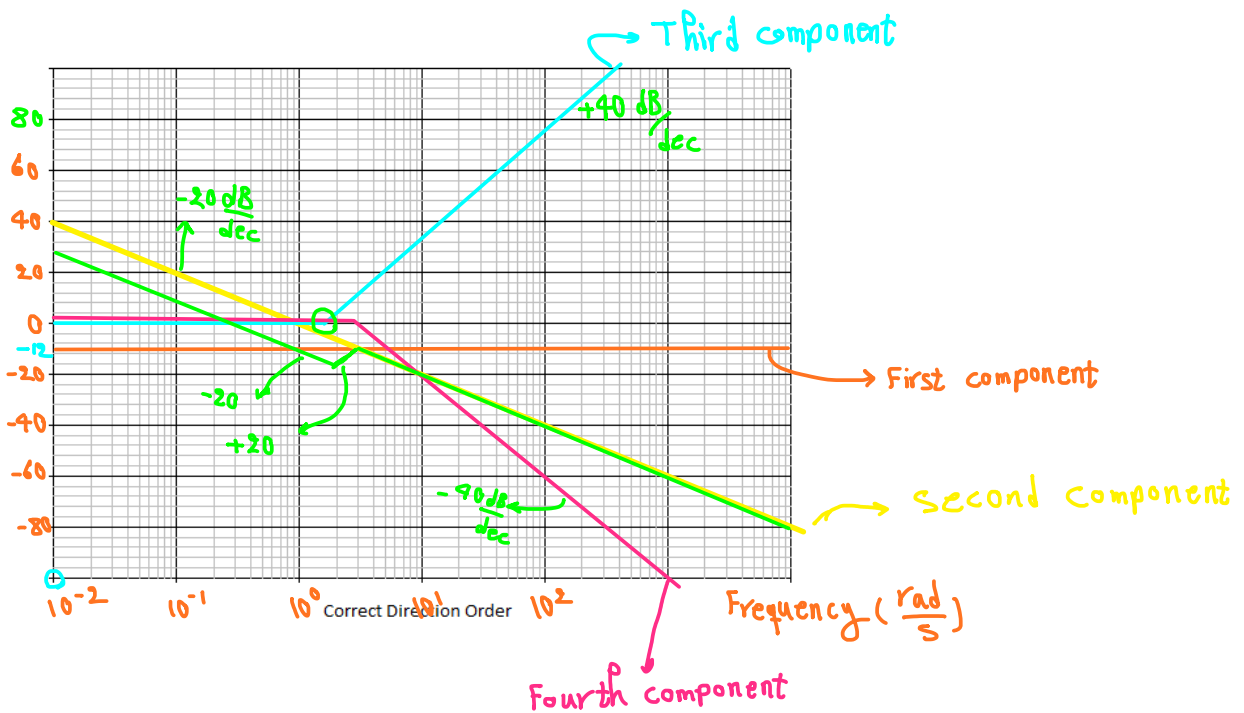
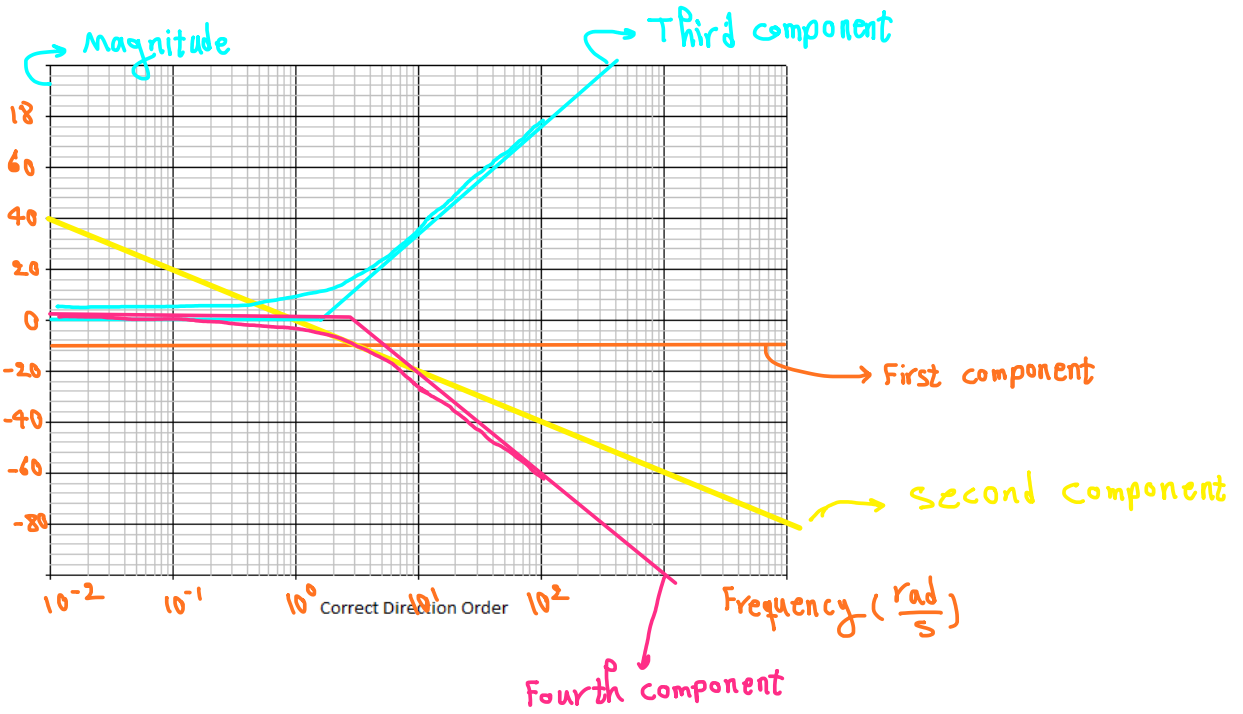
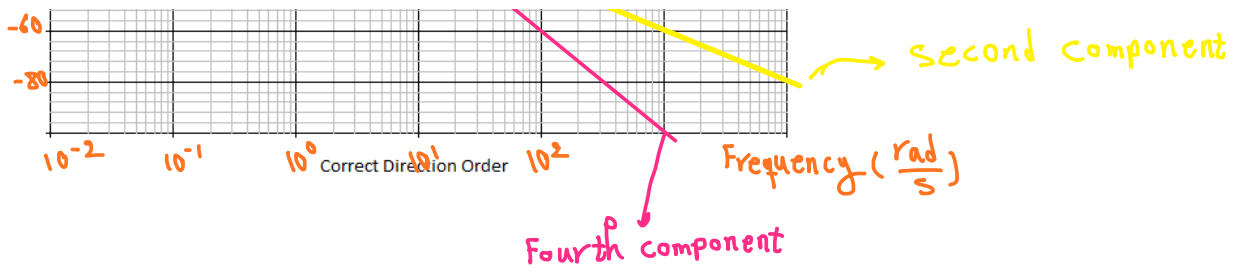
4: The fourth component:  $\frac{1}{1 + \frac{s}{2} + \frac{s^2}{8}}$

Corner frequency or Break frequency:  $\sqrt{8} = 2\sqrt{2} = 2.8284$

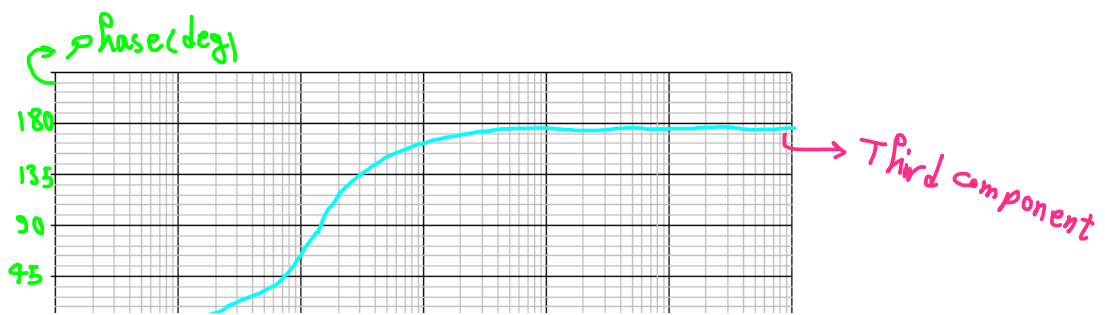
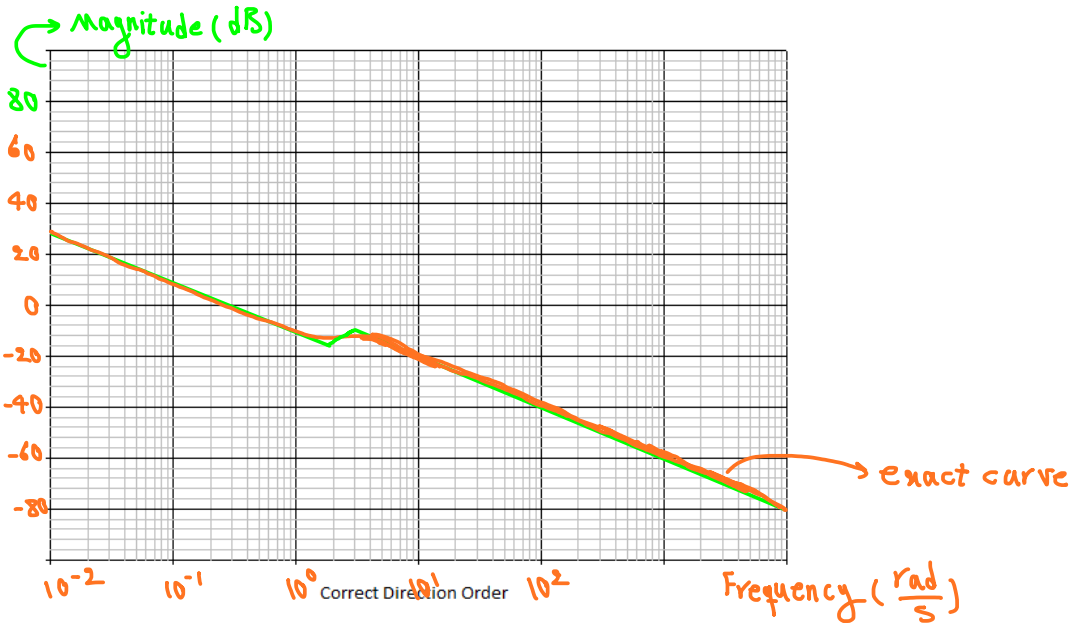
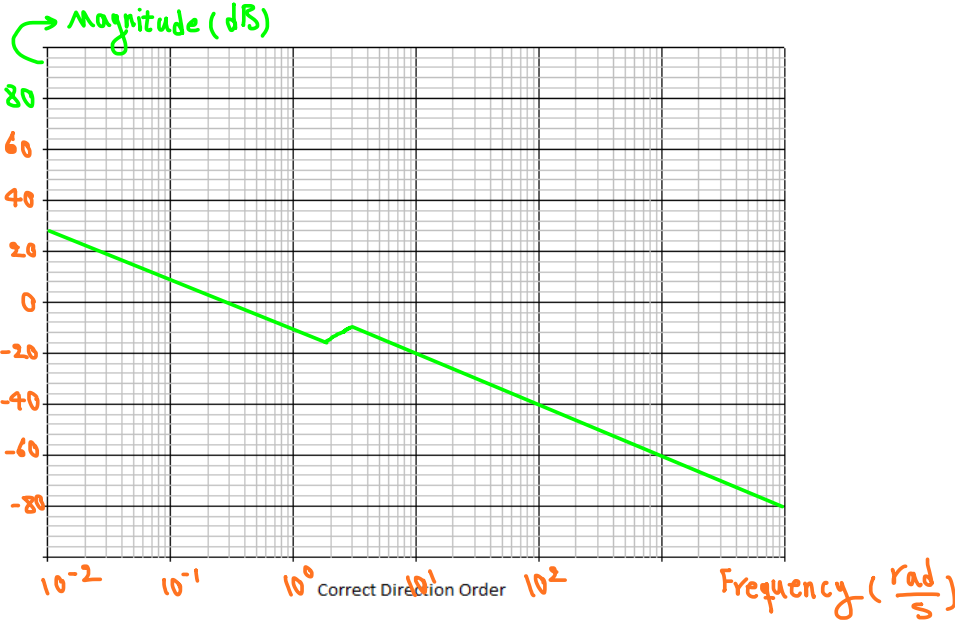


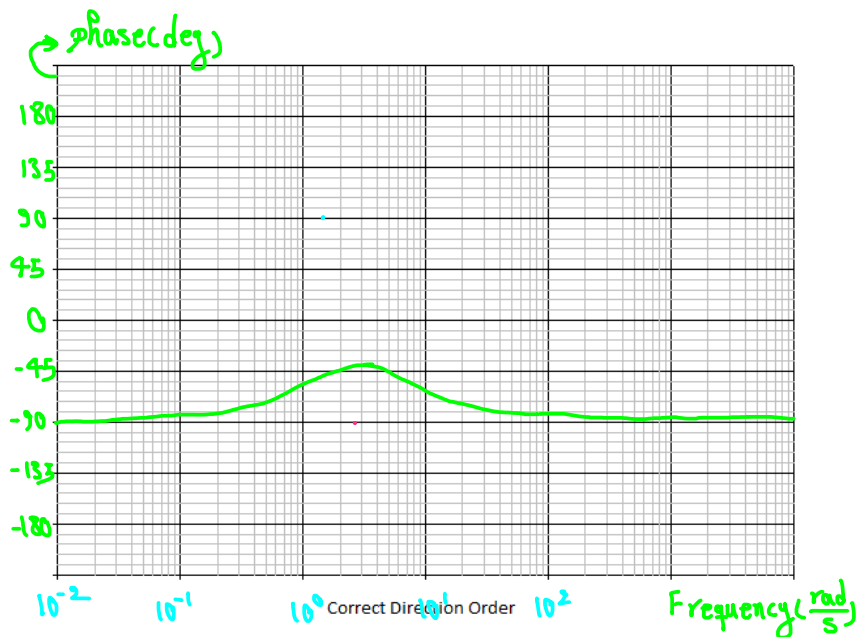
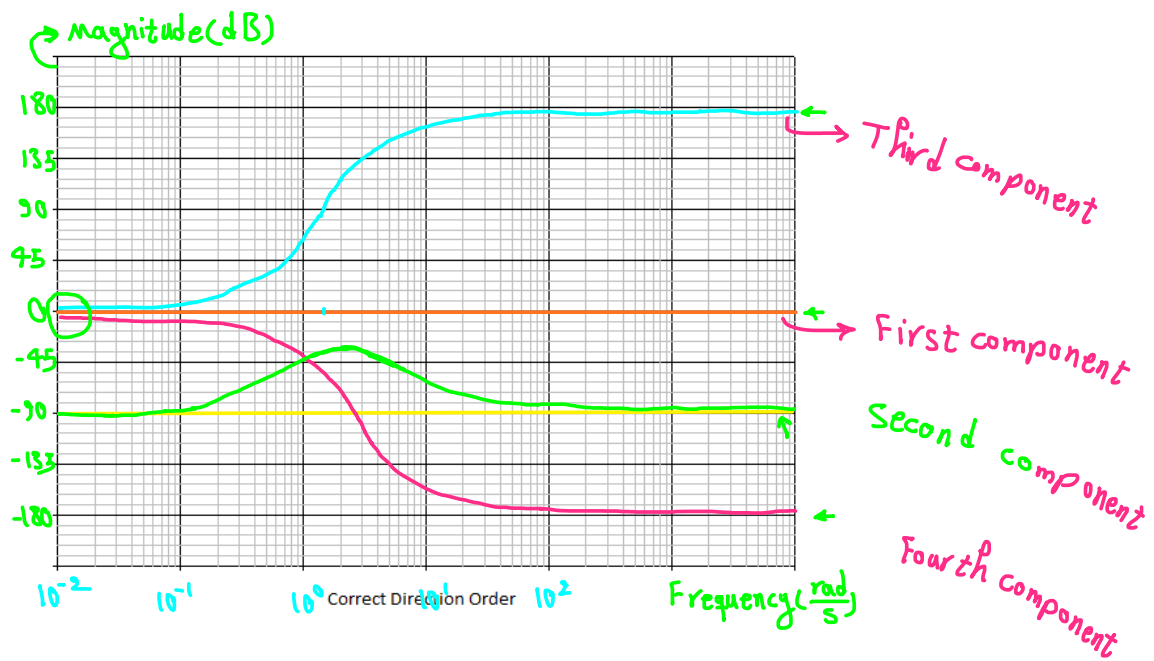
C.F. or B.F. =  $2\sqrt{2}$





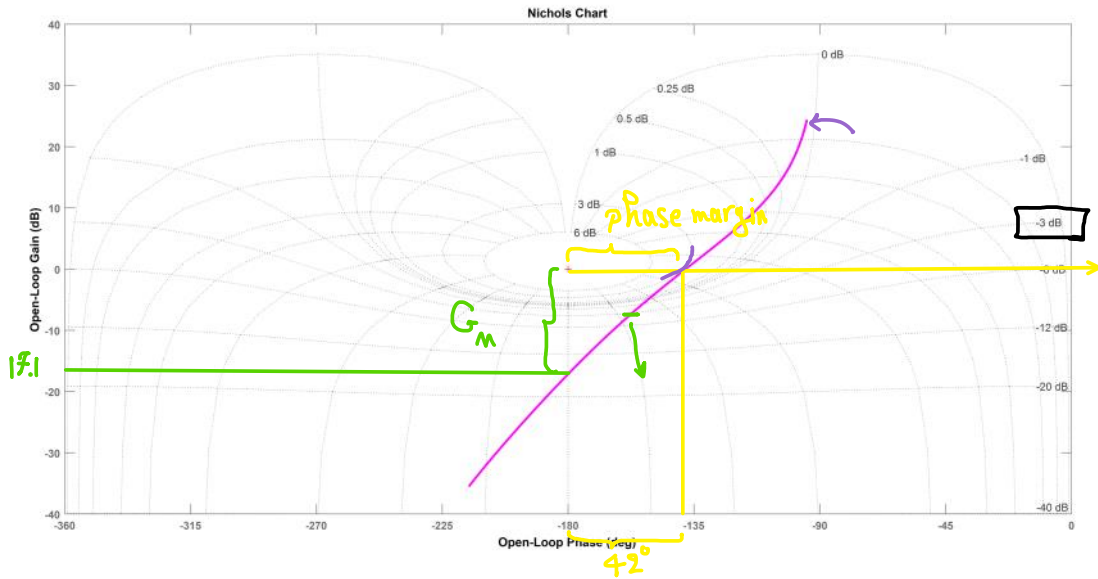
→ Magnitude (dB)







Find Gain margin, Phase margin  $M_r$ ,  $\xi = ?$



$$M_r \approx 2.9 \text{ dB or } 20 \log M_r = 2.9 \Rightarrow M_r = 10^{\left(\frac{2.9}{20}\right)} \approx 1.396$$

$\xi = ?$

Damping ratio ( $\xi$ ) cannot be calculated. Because the order of the system is not stated.

★  $M_p$  is different from  $M_r$